







## **RTQ2551A**

# 1A Ultra-Low Dropout Voltage LDO Regulators with Soft-Start

## 1 General Description

The RTQ2551A is a very low dropout linear regulator that operates from an input voltage as low as 0.8V. The device is capable of supplying 1A of output current with a typical dropout voltage of only 50mV. A VBIAS supply is required to run the internal reference and LDO circuitry while output current comes directly from the VIN supply for high efficiency regulation. User-programmable soft-start limits the input inrush current and minimizes stress on the input power. The enable input and power-good output allow easy sequencing with external regulators. This complete flexibility provides an easy-to-use robust power management solution for a wide variety of applications.

The RTQ2551A is stable with an output capacitor greater than or equal to  $2.2\mu F$ . A precise reference and error amplifier deliver 1% accuracy over load, line and temperature. Overcurrent limit and over-temperature protection are also included. The RTQ2551A is available in the WDFN-10L 3x3 package.

The recommended junction temperature range is  $-40^{\circ}\text{C}$  to  $125^{\circ}\text{C}$ .

### 2 Features

Ultra-Low V<sub>IN</sub> Range: 0.8V to 5.5V
VBIAS Voltage Range: 2.7V to 5.5V
VOUT Voltage Range: 0.8V to 3.6V

Low Dropout: 50mV Typical at 1A, VBIAS = 5V
 1% Accuracy Over Line/Load/ Temperature

Power-Good Indicator for Easy Sequence

Control

 Programmable Soft-Start Provides Linear Voltage Startup

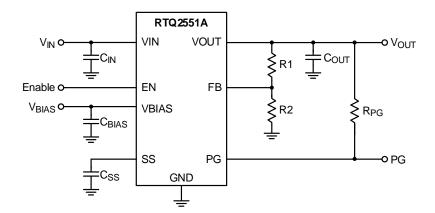
Stable with Any Output Capacitor ≥ 2.2μF

• Overcurrent and Over-Temperature Protection

## 3 Applications

- PCs, Servers, Modems, and Set-Top Boxes
- FPGA Applications
- DSP Core and I/O Voltages
- Instrumentation
- Post-Regulation Applications
- Applications With Sequencing Requirements

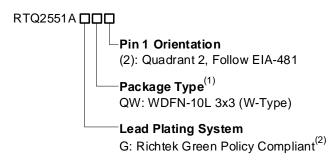
# 4 Simplified Application Circuit



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# 5 Ordering Information



# **6 Marking Information**

W8=: Product Code
YMDAN: Date Code
OAN

#### Note 1.

- Marked with <sup>(1)</sup> indicated: Compatible with the current requirements of IPC/JEDEC J-STD-020.
- Marked with <sup>(2)</sup> indicated: Richtek products are Richtek Green Policy compliant.



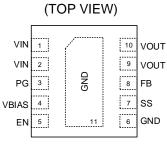
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# 7 Pin Configuration

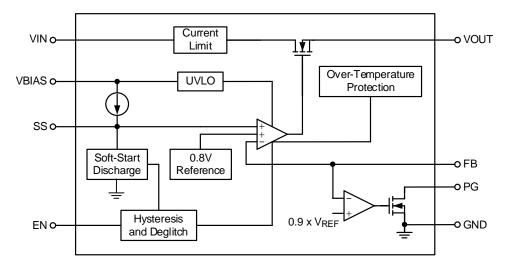


WDFN-10L 3x3

# **8 Functional Pin Description**

Pin No.	Pin Name	Pin Function
1, 2	VIN	Power input of the device.
9, 10	VOUT	Regulated output voltage. A minimum of $2.2\mu\text{F}$ capacitor should be placed directly at this pin.
3	PG	Power-good indicator. An open-drain, active-high output that indicates the status of VOUT. A pull-up resistor from $10k\Omega$ to $1M\Omega$ should be connected from this pin to a supply of up to 5.5V.
4	VBIAS	Bias input pin. Providing input voltage for internal control circuitry.
5	EN	Chip enable (Active-High). Pulling this pin below 0.4V turns the regulator off, reducing the quiescent current to a fraction of its operating value. Connect to VIN if not being used.
6, 11 (Exposed Pad)	GND	Ground. The Exposed Pad must be soldered to a large PCB and connected to GND for maximum power dissipation.
7	SS	Connect a capacitor between this pin and the ground to set the soft-start ramp time of the output voltage.
8	FB	Feedback pin. Connect this pin to an external voltage divider to set the output voltage.

# 9 Functional Block Diagram





## 10 Absolute Maximum Ratings

#### (Note 2)

• Supply Input Voltage, VIN	-0.3V to $6V$
• Other Pins	-0.3V to 6V
Output Voltage, VOUT	-0.3V to $6.3V$
• Lead Temperature (Soldering, 10 sec.)	260°C
• Junction Temperature	150°C
Storage Temperature Range	-65°C to 150°C

Note 2. Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

## 11 ESD Ratings

#### (Note 3)

ESD Susceptibility

HBM (Human Body Model) ----- 2kV

Note 3. Devices are ESD sensitive. Handling precautions are recommended.

## 12 Recommended Operating Conditions

## (Note 4)

• Supply Input Voltage, VIN ----- 0.8V to 5.5V

Note 4. The device is not guaranteed to function outside its operating conditions.

## 13 Thermal Information

#### (Note 5 and Note 6)

	Thermal Parameter	WDFN-10L 3x3	Unit
θЈА	Junction-to-ambient thermal resistance (JEDEC standard)	40.4	°C/W
$\theta$ JC(Top)	Junction-to-case (top) thermal resistance	70.4	°C/W
$\theta$ JC(Bottom)	Junction-to-case (bottom) thermal resistance	13.6	°C/W
θJA(EVB)	Junction-to-ambient thermal resistance (specific EVB)	41.5	°C/W
ΨJC(Top)	Junction-to-top characterization parameter	1.5	°C/W
ΨЈВ	Junction-to-board characterization parameter	25.2	°C/W

Note 5. For more information about thermal parameter, see the Application and Definition of Thermal Resistances report, AN061.

Note 6.  $\theta_{JA(EVB)}$ ,  $\Psi_{JC(TOP)}$  and  $\Psi_{JB}$  are measured on a high effective-thermal-conductivity four-layer test board with dimensions of 70mm x 50mm; furthermore, all layers with 1 oz. Cu. Thermal resistance/parameter values may vary depending on the PCB material, layout, and test environmental conditions.

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## 14 Electrical Characteristics

 $(V_{EN} = 1.1V, V_{IN} = V_{OUT} + 0.3V, V_{BIAS} = 5V, C_{BIAS} = 0.1 \mu F, C_{IN} = C_{OUT} = 10 \mu F, C_{SS} = 1 n F, I_{OUT} = 50 m A, T_{J} = -40 ^{\circ} C \ to \ 125 ^{\circ} C, T_{OUT} = 10 \mu F, T_{OUT} = 10 \mu$ otherwise specified. Typical values are at  $T_A = 25$ °C).

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit	
Input Voltage	VIN		Vout + Vdrop		5.5	V	
VBIAS Pin Voltage	VBIAS		2.7		5.5	V	
Reference Voltage	VREF	TA = 25°C	0.796	0.8	0.804	V	
Output Voltage Range	Vout	VIN = 5V, IOUT = 1A	VREF		3.6	V	
Output Voltage Accuracy	Vout_acc	$2.97V \le V_{BIAS} \le 5.5V$ , $50mA \le I_{OUT} \le 1A$	-1	±0.5	1	%	
Line Regulation	VLINE_REG	VOUT (Normal) + $0.3 \le VIN \le 5.5V$		0.03		%/V	
Load Regulation	VLOAD_REG	50mA ≤ IOUT ≤ 1A		0.09		%/A	
VIN Dropout Voltage	VVIN_DROP	IOUT = 1A, VBIAS - VOUT (Normal) ≥ 3.25V		50	80	mV	
VBIAS Dropout Voltage	Mypiae ppop	IOUT = 1A, VIN = VBIAS		ŀ	1.2	V	
VBIAS Diopout Voltage	VVBIAS_DROP	IOUT = 0.5A, VIN = VBIAS		1	1.1	V	
Current Limit	ILIM	Vout = 80% × Vout (Normal)		1.6		Α	
Bias Pin Current	IBIAS			1	2	mA	
Shutdown Supply Current (IGND)	ISHDN	VEN = 0.4V		1	50	μΑ	
Feedback Pin Current	IFB		-1	0.15	1	μА	
Power-Supply Rejection	PSRR (Note 7)	1kHz, Iout = 0.5A, VIN = 1.8V, Vout = 1.5V		75		dB	
(VIN to VOUT)		300kHz, Iout = 0.5A, VIN = 1.8V, Vout = 1.5V		30		uБ	
Power-Supply Rejection		1kHz, Iout = 0.5A, VIN = 1.8V, Vout = 1.5V 90				- dB	
(VBIAS to VOUT)		300kHz, Iout = 0.5A, VIN = 1.8V, Vout = 1.5V		40		uБ	
Output Noise Voltage	Vn ( <u>Note 7</u> )	100Hz to 100kHz, IOUT = 0.5A, Css = 1nF		25 x Vout		μVRMS	
Minimum Startup Time	tstr (Note 7)	RLOAD for IOUT = 1A, Css = open		200		μs	
Soft-Start Charging Current	Iss	Vss = 0.4V		440		nA	
EN Input Voltage Rising Threshold	V <sub>EN_R</sub>		1.1		5.5	V	
EN Input Voltage Falling Threshold	V <sub>EN_F</sub>		0		0.4	V	
EN Input Hysteresis	VEN_HYS			50		mV	
EN Input Deglitch Time	VEN_DG			20		μS	
EN Input Current	IEN	VEN = 5V		0.1	1	μΑ	
Power-Good Voltage Threshold	VPG	Vout decreasing	85	90	94	%Vout	

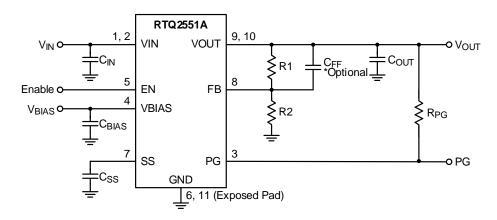


Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Power-Good Voltage Hysteresis	VPG_HYS			3	1	%Vout
Power-Good Output Low Voltage	VPG_L	IPG = 1mA(sinking), VOUT < VPG			0.3	V
Power-Good Leakage Current	VPG_LK	VPG = 5.25V, VOUT > VPG		0.1	1	μΑ
Over-Temperature	Тотр	Shutdown, temperature increasing		165		°C
Protection Threshold		Reset, temperature decreasing		140		

Note 7. Guaranteed by design.



# 15 Typical Application Circuit



<sup>\*:</sup> The feedforward capacitor is optional for the transient response and circuit stability improvement.

**Table 1. Suggested Component Value** 

Vout(V)	R1 (kΩ)	R2 (kΩ)
0.8	Short	Open
0.9	0.619	4.99
1.0	1.13	4.52
1.05	1.37	4.42
1.1	1.87	4.99
1.2	2.49	4.99
1.5	4.12	4.75
1.8	3.57	2.87
2.5	3.57	1.69
3.3	3.57	1.15

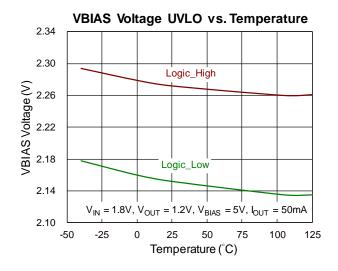
**Table 2. Recommended External Components** 

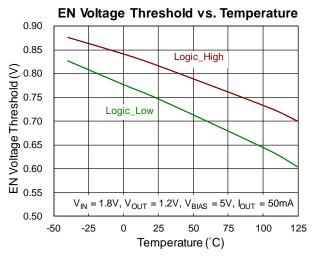
Component	Description	Vendor P/N			
C <sub>IN</sub> , C <sub>OUT</sub> (Note 8)	10μF, 16V, X7S, 0805	GCM21BC71C106KE36 (Murata)			
Css	1nF, 50V, X7R, 0603	GCD188R71H102KA01 (Murata)			
CBIAS	0.1μF, 50V, X7R, 0603	GCJ188R71H104KA12 (Murata)			

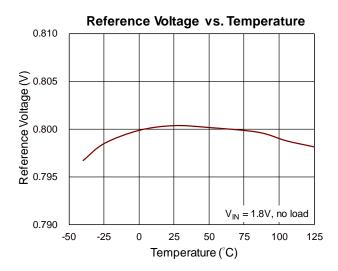
Note 8. Considering the effective capacitance derated with biased voltage level, the Cour component needs satisfy the effective capacitance at least  $2.2\mu F$  or above at targeted output level for stable and normal operation.

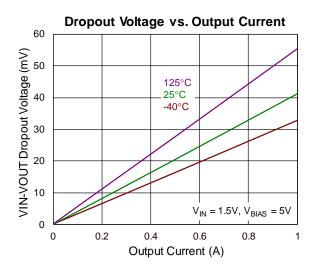


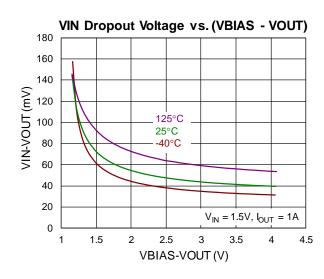
## 16 Typical Operating Characteristics

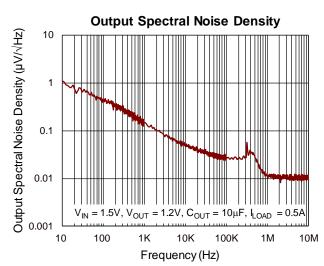






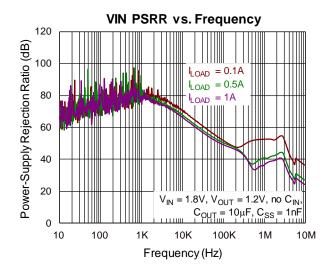


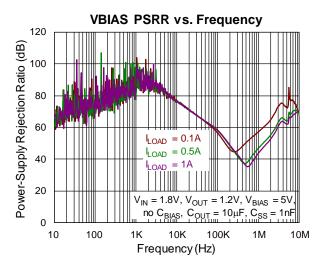




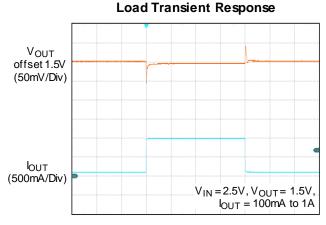
DSQ2551A DS-03







# Power Up Response $C_{SS} = 0nF$ $C_{SS} = 1nF$ $C_{SS} = 2.2nF$ $V_{OUT}$ (500mV/Div) ΕN (2V/Div) $V_{IN} = 1.8V, V_{OUT} = 1.5V, \\ I_{OUT} = 0A, C_{OUT} = 10 \mu F$



Time (1ms/Div) Time (50µs/Div)

10



## 17 Operation

The RTQ2551A is a very low dropout linear regulator that operates from an input voltage as low as 0.8V. It provides a highly accurate output that is capable of supplying 1A of output current with a typical dropout voltage of only 50mV. The output voltage range is from 0.8V to 3.6V.

### 17.1 VIN and VBIAS Supply

The VBIAS input supplies the internal reference and LDO circuitry while all output current comes directly from the VIN input for high efficiency regulation. With an external VBIAS 3.25V above VOUT, the RTQ2551A offers very low dropout performance (80mV Maximum at 1A), which allows the device to be used in place of a DC-DC converter and still achieves good efficiency. This provides designers to achieve the smallest, simplest, and lowest cost solution.

For applications where an auxiliary bias voltage is not available or low dropout is not required. In these applications, VBIAS is suggested to be 1.2V above Vout and attention to power rating and thermal management is needed.

#### 17.2 Enable and Shutdown

The EN pin is active high. Applying a voltage above 1.1V ensures the LDO regulator turns on, while the regulator turns off if the V<sub>EN</sub> falls below 0.4V. The enable circuitry has a typical 50mV hysteresis and deglitching for use with relatively slowly ramping analog signals. That helps avoid on-off cycling as a result of small glitches in the V<sub>EN</sub> signal. A fast rise-time signal must be used to enable the RTQ2551A if precise turn-on timing is required. If not used, EN can be connected to either the VIN or VBIAS pins. If EN is connected to the VIN pin, it should be connected as close as possible to the largest capacitance on the input to prevent voltage droops on that line from triggering the enable circuit.

#### 17.3 Soft-Start

The RTQ2551A includes a soft-start feature to prevent excessive current flow during start-up. When the LDO is enabled, an internal soft-start current (Iss) charges the external soft-start capacitor (Css) to build a ramp-up voltage internally. The RTQ2551A achieves a linear and monotonic soft-start by tracking the voltage ramp until the voltage exceeds the internal reference. The soft-start ramp time can be calculated using the following Equation 1:

$$t_{SS}(ms) = \frac{V_{REF} \times C_{SS}}{I_{SS}} = \frac{0.8V \times C_{SS}(nF)}{0.44\mu A}$$
 (1)

## 17.4 Power-Good Indicator

When the output voltage is greater than VPG + VPG\_HYS, the output voltage is considered good and the open-drain PG pin goes high impedance and is typically pulled high with an external resistor. If Vout drops below VPG or if VBIAS drops below 1.9 V, the open-drain output turns on and pulls the PG output low. The PG pin also asserts when the device is disabled, or when OCP or OTP is triggered.

#### 17.5 Overcurrent Protection

The RTQ2551A has built-in overcurrent protection. When overcurrent (typically 1.6A) is detected, the RTQ2551A starts foldback and limits the current at typically 1.2A. It allows the device to supply surges of up to 1.6A and prevent the device over-heating if a short circuit occurs.



### 17.6 Thermal Shutdown and Over-Temperature Protection

The RTQ2551A includes an over-temperature protection (OTP) circuitry to prevent overheating due to excessive power dissipation. The LDO will shut down when the junction temperature exceeds approximately 165°C. It will re-enable the LDO once the junction temperature drops back to approximately 140°C. The RTQ2551A will cycle in and out of thermal shutdown without latch-up or damage until the overstress condition is removed. Long term overstress (TJ > 125°C) should be avoided as it can degrade the performance or shorten the life of the part.

Note that the over-temperature protection is intended to protect the device during momentary overload conditions. The protection is activated outside the absolute maximum range of operation as a secondary fail-safe and therefore should not be relied upon operationally. Continuous operation above the specified absolute maximum operating junction temperature may impair device reliability or permanently damage the device.

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## 18 Application Information

(Note 9)

The RTQ2551A is a low dropout regulator that features soft-start capability. It provides EN and PG for easily system sequence control, and built-in overcurrent and over-temperature protection for safe operation.

### 18.1 Dropout Voltage

Because of two power supply inputs VIN and VBIAS and one VOUT regulator output, there are two specified dropout voltages. The first is the VIN dropout voltage, which is the voltage difference (VIN - VOUT) when VOUT starts to decrease by percentage specified in the Electrical Characteristics table.

The second, is the VBIAS dropout voltage, which is the voltage difference (VBIAS - VOUT) when the VIN and VBIAS pins are joined together and Vout starts to decrease. This option allows the device to be used in applications where an auxiliary bias voltage is not available or low dropout is not required. In these applications, VBIAS is suggested to be 1.2V above Vout and attention to power rating and thermal considerations is needed.

### 18.2 Input, Output, and Bias Capacitor Selection

The device is designed to be stable for all available types and values of output capacitors  $\geq 2.2\mu F$ . The device is also stable with multiple capacitors in parallel, which can be of any type or value. The capacitance required on the VIN and VBIAS pins strongly depends on the input supply source impedance. To counteract any inductance in the input, the minimum recommended capacitor for VIN is 1µF and the minimum recommended capacitor for VBIAS is 0.1µF. If the VIN and VBIAS pins are connected to the same supply, the recommended minimum capacitor for VBIAS is 4.7μF. Good quality, low ESR capacitors should be used on the input; ceramic X5R and X7R capacitors are preferred. These capacitors should be placed as close to the pins as possible for optimum performance.

#### 18.3 Adjustable the Output Voltage

The output voltage of the RTQ2551A is adjustable from 0.8V to 3.6V using external voltage divider resisters as shown in Typical Application Circuit. R1 and R2 can be calculated to set the desired output voltage. To achieve the maximum accuracy specifications, R2 should be  $\leq 4.99k\Omega$ .

### 18.4 Power Up Sequence Requirement

The RTQ2551A supports powering on the input VIN, VBIAS, and EN pins in any order without damaging the device. Generally, connecting the EN and VIN pins for most applications is acceptable, as long as VIN and VEN are greater than the EN threshold (Minimum = 1.1V) and the input ramp rate of VIN and VBIAS is faster than the output settled soft-start ramp rate. If the VIN/VBIAS input source ramp rate is slower than the output settled soft-start time, the output will track the input supply ramp-up level minus the dropout voltage until it reaches the settled output voltage level. For the other case, if EN is connected with the VBIAS pin, and the provided VIN is present before VBIAS, the output soft-start will proceed as programmed. While VBIAS and VEN are present before VIN is applied and the settled soft-start time has expired, then Vout tracks the Vin ramp-up. If the soft-start time has not expired, the output tracks the VIN ramp-up until the output reaches the value set by the charging soft-start capacitor.



#### 18.5 Thermal Consideration

The junction temperature should never exceed the absolute maximum junction temperature T<sub>J(MAX)</sub>, listed under Absolute Maximum Ratings, to avoid permanent damage to the device. The maximum allowable power dissipation depends on the thermal resistance of the IC package, the PCB layout, the rate of surrounding airflow, and the difference between the junction and ambient temperatures. The maximum power dissipation can be calculated using the following formula:

$$PD(MAX) = (TJ(MAX) - TA) / \theta JA$$

where  $T_{J(MAX)}$  is the maximum junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction-to-ambient thermal resistance.

For continuous operation, the maximum operating junction temperature indicated under Recommended Operating Conditions is 125°C. The junction-to-ambient thermal resistance,  $\theta_{JA(EVB)}$ , is highly package dependent. For a WDFN-10L 3x3 package, the thermal resistance,  $\theta_{JA(EVB)}$ , is 41.5°C/W on a high effective-thermal-conductivity four- layer test board. The maximum power dissipation at  $T_A = 25$ °C can be calculated as below:

$$P_{D(MAX)} = (125^{\circ}C - 25^{\circ}C) / (41.5^{\circ}C/W) = 2.41W$$
 for a WDFN-10L 3x3 package.

The maximum power dissipation depends on the operating ambient temperature for the fixed  $T_{J(MAX)}$  and the thermal resistance,  $\theta_{JA(EVB)}$ . The derating curve in <u>Figure 1</u> allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.

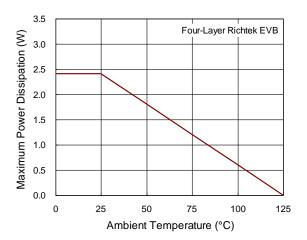


Figure 1. Derating Curve of Maximum Power Dissipation

#### 18.6 Layout Considerations

For best performance of the RTQ2551A, the following PCB layout suggestions are highly recommended:

- The input capacitor must be placed as close as possible to the IC to minimize the power loop area.
- Minimize the power trace length and avoid using vias for the input and output capacitors connection.

<u>Figure 2</u> shows the example for the layout reference that helps minimize inductive parasitic components, reduce load transients, and ensure good circuit stability.



Wider for thermal considerations.

GND Plane

Vout Plane

Vout Plane

Vout Plane

Vout Plane

Vin 13 0 0 15 Vout Vout 15 Vout Vin 23 0 0 15 Vout 15 Vo

The GND layout trace should be

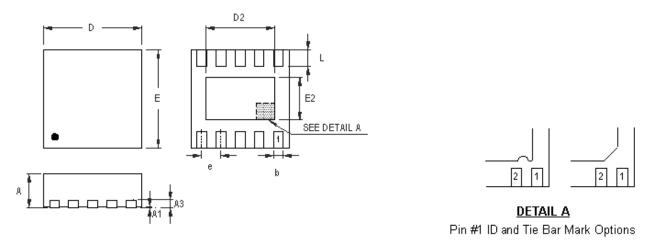
Figure 2. PCB Layout Guide

**Note 9.** The information provided in this section is for reference only. The customer is solely responsible for the designing, validating, and testing your product incorporating Richtek's product and ensure such product meets applicable standards and any safety, security, or other requirements.

RICHTEK is a



## 19 Outline Dimension



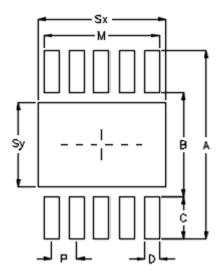
Note: The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

Comple of	Dimensions I	n Millimeters	Dimensions In Inches		
Symbol	Min	Max	Min	Max	
Α	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
A3	0.175	0.250	0.007	0.010	
b	0.180	0.300	0.007	0.012	
D	2.950	3.050	0.116	0.120	
D2	2.300	2.650	0.091	0.104	
Е	2.950	3.050	0.116	0.120	
E2	1.500	1.750	0.059	0.069	
е	0.500		0.0	)20	
L	0.350	0.450	0.014	0.018	

W-Type 10L DFN 3x3 Package



# 20 Footprint Information

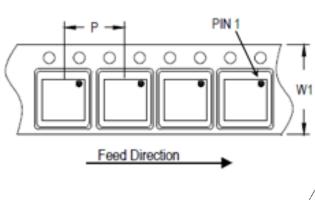


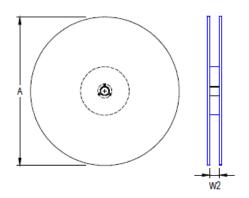
Package	Number of	Footprint Dimension (mm)						Tolerance		
	Pin	Р	Α	В	С	D	Sx	Sy	М	
V/W/U/X/ZDFN3*3-10	10	0.50	3.80	2.10	0.85	0.30	2.55	1.70	2.30	±0.05

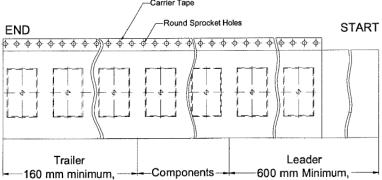


# 21 Packing Information

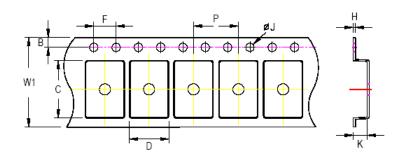
## 21.1 Tape and Reel Data







Package Type	Tape Size (W1) (mm)	Pocket Pitch (P) (mm)	Reel S (mm)	ize (A)	Units per Reel	Trailer (mm)	Leader (mm)	Reel Width (W2) Min./Max. (mm)
QFN/DFN 3x3	12	8	180	7	1,500	160	600	12.4/14.4



- C, D, and K are determined by component size. The clearance between the components and the cavity is as follows:
- For 12mm carrier tape: 0.5mm max.

Tape Size	W1	Р		В		F		ØJ		Н
	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Min.	Max.	Max.
12mm	12.3mm	7.9mm	8.1mm	1.65mm	1.85mm	3.9mm	4.1mm	1.5mm	1.6mm	0.6mm

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## 21.2 Tape and Reel Packing

Step	Photo/Description	Step	Photo/Description
1	Reel 7"	4	RICHTER CARRY TO THE PROPERTY OF THE PROPERTY
2	HIC & Desiccant (1 Unit) inside	5	12 inner boxes per outer box
3	Caution label is on backside of Al bag	6	Outer box Carton A

Container	Reel		Вох			Carton		
Package	Size	Units	Item	Reels	Units	Item	Boxes	Unit
OFN/DEN 00	7"	1,500	Box A	3	4,500	Carton A	12	54,000
QFN/DFN 3x3			Box E	1	1,500	For Combined or Partial Reel.		Reel.

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## 21.3 Packing Material Anti-ESD Property

Surface Resistance	Aluminum Bag	Reel	Cover tape	Carrier tape	Tube	Protection Band
$\Omega$ /cm <sup>2</sup>	10 <sup>4</sup> to 10 <sup>11</sup>					

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22 Datasheet Revision History

Version	Date	Description	Item
00	2023/12/6	Final	Electrical Characteristics on page 7 Operation on page 12
01	2024/3/8	Modify	General Description on page 1 Features on page 1 Electrical Characteristics on page 7 Operation on page 12, 13 Application Information on page 14
02	2024/7/1	Modify	Ordering Information on page 2 Electrical Characteristics on page 7, 8 Packing Information on page 20
03	2024/10/7	Modify	Changed the names of pin 3 to PG. Simplified Application Circuit on page 1 Ordering Information on page 2 Functional Block Diagram on page 4 Electrical Characteristics on page 6, 7 Typical Application Circuit on page 8 Operation on page 11, 12 Application Information on page 13, 15