

# **Source/Sink DDR Termination Regulator**

### **General Description**

The RT9066 is a source/sink tracking termination regulator. It is specifically designed for low-cost and low-external component count systems. The RT9066 possesses a high speed operating amplifier that provides fast load transient response and only requires a minimum  $30\mu F$  of ceramic output capacitance. The RT9066 supports remote sensing functions and all features required to power the DDRII / DDRIII and Low Power DDRIII / DDRIV VTT bus termination according to the JEDEC specification. In addition, the RT9066 provides an open drain PGOOD signal to monitor the output regulation and an EN signal that can be used to discharge VTT during S3 (suspend to RAM) for DDR applications .

The RT9066 is available in the thermal efficient package WDFN-16L 5x3.

## **Marking Information**

01=YM DNN 01= : Product Code YMDNN : Date Code

### **Features**

• VIN Input Voltage Range: 1.1V to 2V

VCNTL Input Voltage Range: 2.375V to 5.5V

• Output Current Up to 4.5A

MLCC Stable

• PGOOD to Monitor Output Regulation

• ±10mA Reference (REFOUT)

 Meet DDRII JEDEC Spec and Support DDRIII, Low Power DDRIII / DDRIV VTT Application

Soft-Start Function

UVLO and OCP Protection

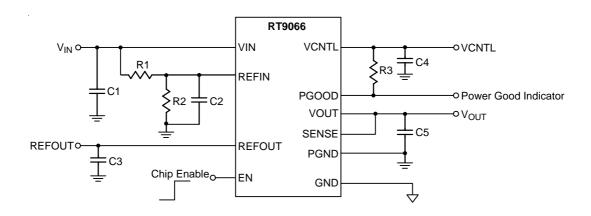
• Thermal Shutdown

• RoHS Compliant and Halogen Free

### **Applications**

- Notebook/Desktop/Server
- Telecom/Datacom, GSM Base Station, LCD-TV/PDP-TV ,Copier/Printer, Set-Top Box

# **Simplified Application Circuit**



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# **Ordering Information**

### RT9066 □□ Package Type QW: WDFN-16L 5x3 (W-Type) Lead Plating System G: Green (Halogen Free and Pb Free)

#### Note:

Richtek products are:

- ▶ RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- ▶ Suitable for use in SnPb or Pb-free soldering processes.

# **Pin Configurations**

(TOP VIEW) GND 1 16 GND REFIN 2 VIN 3 VOUT 4 PGND 5 SENSE 6 GND 7 **VCNTL** PGOOD GND GND 12 EN 11 REFO 10 GND REFOUT GND 8 9 GND

WDFN-16L 5x3

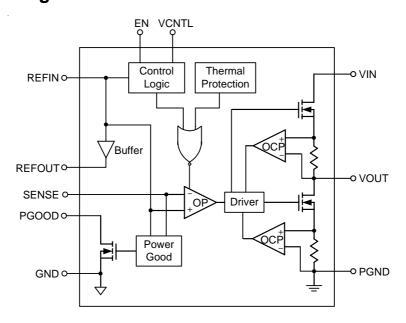
## **Functional Pin Description**

Pin No.	Pin Name	Pin Function
1, 7, 8, 9, 10, 13, 16, 17 (Exposed Pad)	GND	Signal Ground. The exposed pad must be soldered to a large PCB and connected to GND for maximum power dissipation.
2	REFIN	Reference Voltage Input.
3	VIN	Supply Voltage Input.
4	VOUT	Power Output.
5	PGND	Power Ground.
6	SENSE	Voltage Sense Input. Connect to positive terminal of the output capacitor or the load.
11	REFOUT	Reference Voltage Output. Connect a 0.1μF ceramic capacitor to GND.
12	EN	Chip Enable. For DDR VTT application, connect EN to SLP_S3. For any other application(s), use EN as the ON/OFF function.
14	PGOOD	PGOOD Open Drain Output. Connect a $100 k\Omega$ pull-high resistor to VCNTL.
15	VCNTL	2.5V, 3.3V or 5V Power Supply. A ceramic decoupling capacitor with a value between $1\mu F$ and $4.7\mu F$ is required.

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### **Function Block Diagram**



## **Operation**

The RT9066 is a linear sink/source DDR termination regulator with current capability up to 4.5A. The RT9066 builds in a high side N-type power MOSFET which provides current sourcing and a low side N-type power MOSFET which provides current sinking. All the control circuits are supplied by the power VCNTL. In normal operation, the error amplifier OP adjusts the gate driving voltage of the power MOSFET to achieve SENSE voltage well tracking the REFIN voltage.

Both the source and sink currents are detected by the internal sensing resistor, and the OCP function will work to limit the current to a designed value when overload happens. Furthermore, the current will be folded back to be one half if VOUT is out of the power good window.

#### **Buffer**

This function provides REFOUT output equal to REFIN with 10mA source/sink current capability.

#### **Power Good**

When the SENSE voltage is in the power good window and lasts for a certain delay time, the PGOOD pin will be high impedance, and the PGOOD voltage will be pulled high by the external resistor.

#### **Control Logic**

This block includes VCNTL UVLO, REFIN UVLO and Enable/Disable function, provides logic control to the whole chip.

#### **Thermal Protection**

Both the high side and low side power MOSFET will be turned off when the junction temperature is higher than typically 160°C, and released to normal operation when junction temperature falls below typically 120°C.



### **Absolute Maximum Ratings** (Note 1)

Supply Voltage, VIN, VCNTL	-0.3V to 6V
• Input Voltage, EN, REFIN, SENSE	- −0.3V to 6V
Output Voltage, VOUT, REFOUT, PGOOD	-0.3V to 6V
<ul> <li>Power Dissipation, P<sub>D</sub> @ T<sub>A</sub> = 25°C</li> </ul>	
WDFN-16L 5x3	· 3.333W
Package Thermal Resistance (Note 2)	
WDFN-16L 5x3, $\theta_{JA}$	- 30°C/W
WDFN-16L 5x3, $\theta_{JC}$	· 7.5°C/W
• Lead Temperature (Soldering, 10 sec.)	- 260°C
• Junction Temperature	· 150°C
Storage Temperature Range	-65°C to 150°C
ESD Susceptibility (Note 3)	
HBM (Human Body Model)	· 2kV
MM (Machine Model)	- 200V
Recommended Operating Conditions (Note 4)	

• Control Input Voltage, VCNTL ----- 2.375V to 5.5V • Supply Input Voltage, VIN ----- 1.1V to 2V • Junction Temperature Range ----- --- -40°C to 125°C • Ambient Temperature Range ----- --- -40°C to 85°C

### **Electrical Characteristics**

 $(V_{IN} = 1.5V, V_{EN} = V_{CNTL} = 5V, V_{REFIN} = V_{SENSE} = 0.75V, C_{OUT} = 10 \mu F x 5, T_A = 25 ^{\circ}C$ , unless otherwise specified)

Parameter	Symbol	I Test Conditions		Тур	Max	Unit	
Supply Current							
VCNTL Supply Current	Ivcntl	V <sub>EN</sub> = V <sub>CNTL</sub> , No Load		1.5	2	mA	
VCNTL Shutdown Current	I <sub>SHDN_VCNTL</sub>	$V_{EN} = 0V$ , $V_{REFIN} = 0V$ , No Load		65	80	μА	
		$V_{EN} = 0V$ , $V_{REFIN} > 0.4V$ , No Load	-	200	400	μΑ	
VIN Supply Current	I <sub>VIN</sub>	V <sub>EN</sub> = V <sub>CNTL</sub> , No Load	1		10	mA	
VIN Shutdown Current	I <sub>SHDN_VIN</sub>	V <sub>EN</sub> = 0V, No Load		0.1	50	μΑ	
Output							
Output DC Voltage	Vsense	V <sub>IN</sub> = 1.8V, V <sub>REFIN</sub> = 0.9V		0.9		V	
		(DDRII), $I_{OUT} = 0A$	-10		10	mV	
		V <sub>IN</sub> = 1.5V, V <sub>REFIN</sub> = 0.75V		0.75		V	
		(DDRIII), I <sub>OUT</sub> = 0A	-10		10	mV	
Output Voltage Tolerance to REFIN	Vvotol	-4.3A < I <sub>OUT</sub> < 4.3A	-20		20	mV	

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Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit	
VOUT Source Current Limit	I <sub>LIM_</sub> VOUT_SR	VOUT in PGOOD window	5		9	Α	
VOUT Sink Current Limit	I <sub>LIM_</sub> VOUT_SK	VOUT in PGOOD window	5		9	Α	
VOUT Discharge Resistance	RDISCHARGE	$V_{REFIN} = 0V$ , $V_{OUT} = 0.3V$ , $V_{EN} = 0V$		18	25	Ω	
Power Good Comparator							
		V <sub>SENSE</sub> lower threshold with respect to V <sub>REFIN</sub>	-23.5	-20	-17.5		
PGOOD Threshold	V <sub>TH_PGOOD</sub>	V <sub>SENSE</sub> upper threshold with respect to V <sub>REFIN</sub>	17.5	20	23.5		
		PGOOD Hysteresis		5			
PGOOD Start-Up Delay	T <sub>PGDELAY1</sub>	Start-up rising edge, V <sub>SENSE</sub> within 15% of V <sub>REFIN</sub>		2		ms	
Output Low Voltage	V <sub>LOW_PGOOD</sub>	I <sub>SINK</sub> = 4mA			0.2	V	
PGOOD Bad Delay	T <sub>PGDELAY2</sub>	V <sub>SENSE</sub> is outside of ±20% PGOOD window		10		μS	
Leakage Current	I <sub>LEAKAGE _PGOOD</sub>	V <sub>SENSE</sub> = V <sub>REFIN</sub> (PGOOD high impedance), V <sub>PGOOD</sub> = V <sub>CNTL</sub> + 0.2V			1	μА	
REFIN and REFOUT	•		•	•	•		
REFIN Input Current	I <sub>REFIN</sub>	V <sub>EN</sub> = V <sub>CNTL</sub>			1	μА	
REFIN Voltage Range	V <sub>REFIN</sub>		0.5		1.8	V	
REFIN Under Voltage	V	REFIN Rising	360	390	420	m∨	
Lockout	Vuvlo_refin	Hysteresis 20		20		1117	
	VTOL_REFOUT	$-10$ mA < $I_{REFOUT}$ < $10$ mA, $V_{REFIN}$ = $0.9$ V	-15		15	mV	
REFOUT Voltage Tolerance to VREFIN		$-10$ mA < $I_{REFOUT}$ < $10$ mA, $V_{REFIN} = 0.75$ V	-15		15		
		$-10$ mA < $I_{REFOUT}$ < $10$ mA, $V_{REFIN}$ = $0.6$ V	-15		15		
REFOUT Source Current Limit	I <sub>LIM_REFOUT_SR</sub>	V <sub>REFOUT</sub> = 0V	10	40		mA	
REFOUT Sink Current Limit	I <sub>LIM_REFOUT_SK</sub>	V <sub>REFOUT</sub> = V <sub>IN</sub>	10	40		mA	
UVLO/EN Logic Threshold							
UVLO Threshold	V <sub>UVLO_VCNTL</sub>	Rising	2.2	2.3	2.375	V	
OVEO TITICONOIG	VOVEO_VCIVIE	Hysteresis		100		mV	
Enable High-Level Input Voltage	V <sub>IN_H</sub>		1.7			V	
Enable Low-Level Input Voltage	V <sub>IN_L</sub>				0.3	V	



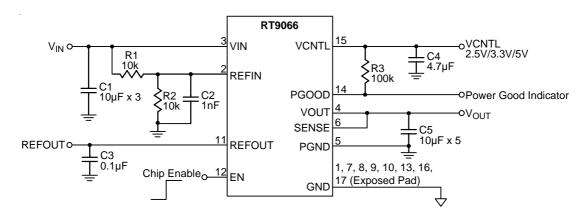
Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
Enable Hysteresis Voltage	V <sub>EN_HYS</sub>			0.5		V
Enable Logic Input Leakage Current	I <sub>LEAKAGE_EN</sub>		-1		1	μА
Thermal Shutdown			•			
Thermal Shutdown	TSD	Shutdown Temperature		160		°C
Threshold		Hysteresis		40		

- Note 1. Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- Note 2.  $\theta_{JA}$  is measured at  $T_A = 25$ °C on a high effective thermal conductivity four-layer test board per JEDEC 51-7.  $\theta_{JC}$  is measured at the exposed pad of the package.
- Note 3. Devices are ESD sensitive. Handling precaution is recommended.
- Note 4. The device is not guaranteed to function outside its operating conditions.

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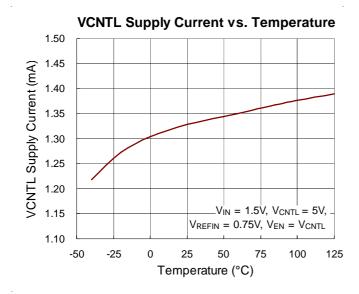


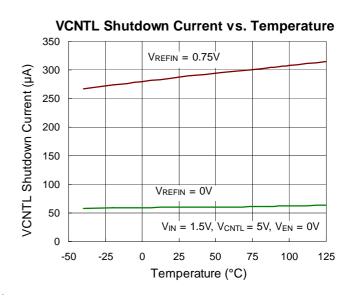
# **Typical Application Circuit**

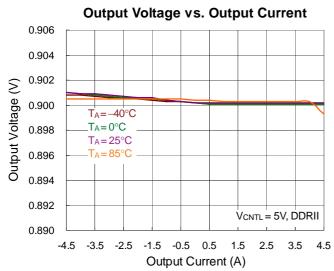


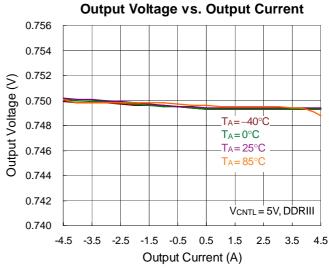


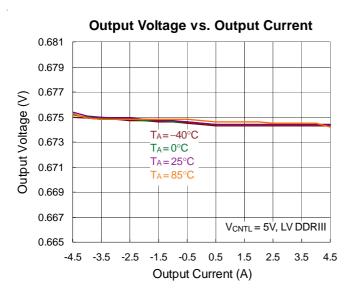
## **Typical Operating Characteristics**

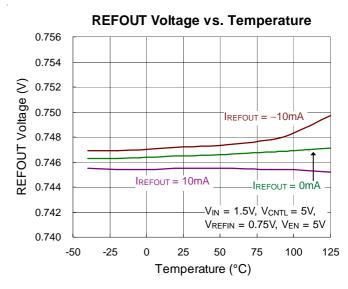




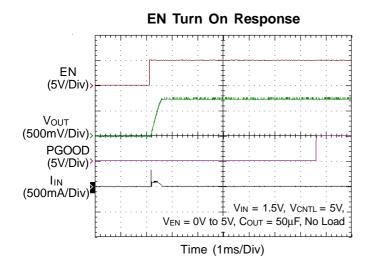


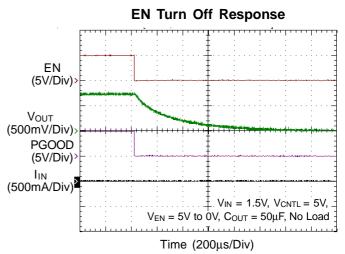


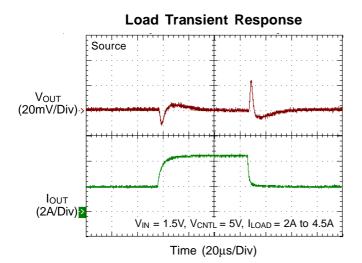


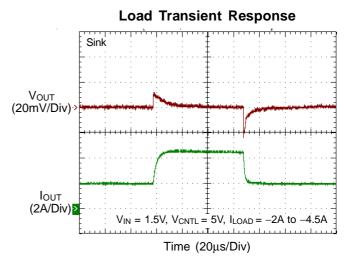














### **Application Information**

The RT9066 is a sink/source tracking termination regulator. It is specifically designed for low-cost and low-external component count systems such as notebook PC applications. The RT9066 possesses a high speed operating amplifier that provides fast load transient response and only requires three 10µF ceramic input capacitors and five 10µF ceramic output capacitors.

#### **REFOUT Regulator**

REFOUT is a reference output voltage with source/sink current capability up to 10mA. To ensure stable operation, a 0.1µF ceramic capacitor connected between REFOUT and GND is recommended.

#### **Capacitor Selection**

To achieve best performance of the RT9066, it is recommended to follow the following descriptions for capacitor selection.

#### **VCNTL Capacitor**

Add a ceramic capacitor 4.7µF placed to VCNTL pin as close as possible to stabilize the supply voltage (2.5V, 3.3V or 5V rail) from any parasitic impedance from the supply.

#### **VIN Capacitor**

Good bypassing is recommended from VIN to GND to improve transient response. It is recommended to place three 10µF or greater input capacitors as close as possible to the IC and the distance must be less than 0.5 inch from the VIN pin.

#### **VOUT Capacitor**

For stable operation, the total capacitance of the VTT output terminal must be greater than 30µF. The RT9066 is designed specifically to work with low ESR ceramic output capacitor in space-saving and performance consideration. Larger output capacitance can reduce the noise and improve load transient response, stability and PSRR. Five 10μF ceramic capacitors are used in the typical application circuit. The output capacitor should be located near the VOUT pin as close as possible.

#### **Operation State Setting**

The EN pin can be connected to SLP S3 signal for DDR VTT application. Both VOUT and REFOUT are turned on in normal state (EN = High, REFIN > 0.39V). In standby state (EN = Low, REFIN > 0.39V), REFOUT voltage is kept alive and VOUT voltage is turned off and discharged via internal MOSFET. When EN = Low and REFIN < 0.39V, the RT9066 enters shutdown state, and VOUT and REFOUT are turned off and discharged to ground via internal MOSFETs. Table 1 summarizes the abovementioned operation state setting, and Figure 1 shows a typical start-up and shutdown timing diagram.

**Table 1. Operation State Setting** 

State	EN	REFIN	VOUT	REFOUT
Normal	High	> 0.39V	On	On
Standby	Low	> 0.39V	Off	On
Shutdown	Low	< 0.39V	Off	Off

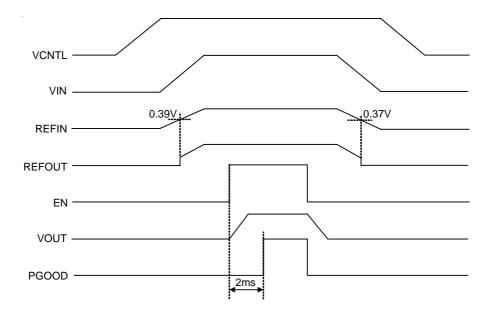


Figure 1. Typical Start-up and Shutdown Timing Diagram

#### **Thermal Considerations**

For continuous operation, do not exceed absolute maximum operation junction temperature. The maximum power dissipation depends on the thermal resistance of IC package, PCB layout, the rate of surroundings airflow and temperature difference between junction to ambient. The maximum power dissipation can be calculated by following formula:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

Where  $T_{J(MAX)}$  is the maximum operation junction temperature,  $T_A$  is the ambient temperature and the  $\theta_{JA}$  is the junction to ambient thermal resistance.

For recommended operating conditions specification, the maximum junction temperature is 125°C. The junction to ambient thermal resistance  $\theta_{JA}$  is layout dependent. For WDFN-16L 5x3 package, the thermal resistance  $\theta_{JA}$  is 30°C/W on the standard JEDEC 51-7 four layers thermal test board. The maximum power dissipation at  $T_A = 25$ °C can be calculated by following formula:

$$P_{D(MAX)} = (125^{\circ}C - 25^{\circ}C) / (30^{\circ}C/W) = 3.333W$$
 for WDFN-16L 5x3

The maximum power dissipation depends on operating ambient temperature for fixed T<sub>J(MAX)</sub> and thermal resistance  $\theta_{JA}$ . The derating curve in Figure 2 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation allowed.

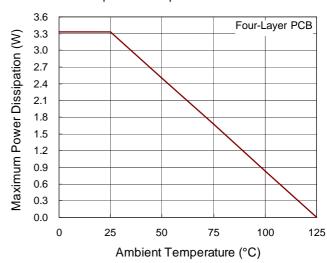
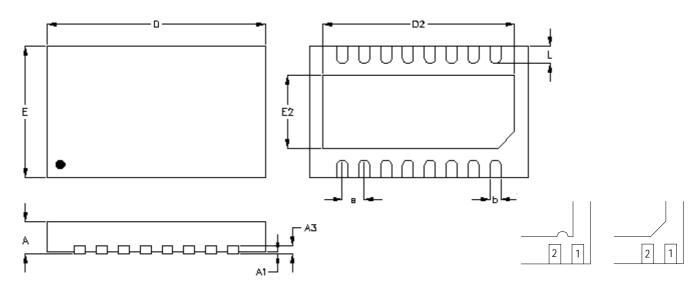


Figure 2. Derating Curve of Maximum Power Dissipation



### **Outline Dimension**



**DETAIL A** 

Pin #1 ID and Tie Bar Mark Options

Note: The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

Symbol	Dimensions I	n Millimeters	Dimensions In Inches		
	Min.	Max.	Min.	Max.	
А	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
А3	0.175	0.250	0.007	0.010	
b	0.180	0.300	0.007	0.012	
D	4.950	5.050	0.195	0.199	
D2	4.350	4.450	0.171	0.175	
E	2.950	3.050	0.116	0.120	
E2	1.600	1.700	0.063	0.067	
е	0.5	500	0.0	20	
L	0.350	0.450	0.014	0.018	

W-Type 16L DFN 5x3 Package

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